



# Chaos Suppression and Stabilization of Generalized Liu Chaotic Control System

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## ABSTRACT

In this paper, the concept of generalized stabilization for nonlinear systems is introduced and the stabilization of the generalized Liu chaotic control system is explored. Based on the time-domain approach with differential inequalities, a suitable control is presented such that the generalized stabilization for a class of Liu chaotic system can be achieved. Meanwhile, not only the guaranteed exponential convergence rate can be arbitrarily pre-specified but also the critical time can be correctly estimated. Finally, some numerical simulations are given to demonstrate the feasibility and effectiveness of the obtained results.

**Key Words:** Generalized synchronization, Liu chaotic system, critical time, exponential convergence rate

## 1. INTRODUCTION

In recent years, chaotic dynamic systems have been widely investigated by researchers; see, for instance, [1-12] and the references therein. Very often, chaos in many dynamic systems is an origin of the generation of oscillation and an origin of instability. For a chaotic control system, it is important to design a controller that has both good transient and steady-state response. Furthermore, suppressing the occurrence of chaos plays an important role in the controller design of a nonlinear system.

In the past decades, various methodologies in control design of chaotic system have been presented, such as variable structure control approach, time-domain approach, adaptive control approach, adaptive sliding mode control approach, back stepping control approach, and others.

In this paper, the concept of generalized stabilizability for nonlinear dynamic systems is introduced and the stabilizability of generalized Liu chaotic control system will be investigated. Based on the time-domain approach with differential inequality, a suitable control will be offered such that the generalized stabilization can be achieved for a class of Liu chaotic system. Not only the critical time can be correctly estimated, but also the guaranteed exponential convergence rate can be arbitrarily pre-specified. Several numerical simulations will also be provided to illustrate the use of the main results.

The layout of the rest of this paper is organized as follows. The problem formulation, main result, and controller design procedure are presented in Section 2. Numerical simulations are given in Section 3 to show the effectiveness of the developed results. Finally, conclusion remarks are drawn in Section 4.

## 2. PROBLEM FORMULATION AND MAIN RESULTS

### Nomenclature

$\mathfrak{R}^n$	the $n$ -dimensional real space
$ a $	the modulus of a complex number $a$
$A^T$	the transport of the matrix $A$
$\ x\ $	the Euclidean norm of the vector $x \in \mathfrak{R}^n$

In this paper, we explore the following generalized Liu chaotic system:

$$\dot{x}_1(t) = a_1 x_1(t) + a_2 x_2(t) + u_1(t), \quad (1a)$$

$$\dot{x}_2(t) = a_3 x_1(t) + a_4 x_1(t)x_3(t) + u_2(t), \quad (1b)$$

$$\dot{x}_3(t) = a_5 x_1(t) + a_6 x_2(t) + a_7 x_3(t) + a_8 x_1^2(t) + u_3(t), \quad \forall t \geq 0, \quad (1c)$$

$$\begin{aligned} & [x_1(0) \ x_2(0) \ x_3(0)]^T \\ & = [x_{10} \ x_{20} \ x_{30}]^T, \end{aligned} \tag{1d}$$

where  $x(t) := [x_1(t) \ x_2(t) \ x_3(t)]^T \in \mathfrak{R}^3$  is the state vector,  $u(t) := [u_1(t) \ u_2(t) \ u_3(t)]^T \in \mathfrak{R}^3$  is the system control,  $[x_{10} \ x_{20} \ x_{30}]^T$  is the initial value, and  $a_i, b \in \mathfrak{R}$  indicate the parameters of the system. The original Liu chaotic system is a special case of system (1) with  $u(t) = 0$ ,  $a_1 = -a_2 = 10$ ,  $a_3 = 40$ ,  $a_4 = -1$ ,  $a_5 = a_6 = 0$ ,  $a_7 = -2.5$ , and  $a_8 = 4$ . It is well known that the system (1) without any control (i.e.,  $u(t) = 0$ ) displays chaotic behavior for certain values of the parameters [1]. The aim of this paper is to search a novel control for the system (1) such that the generalized stability of the feedback-controlled system can be guaranteed. In this paper, the concept of generalized stabilization will be introduced. Motivated by time-domain approach with differential inequality, a suitable control strategy will be established. Our goal is to design a control such that the generalized stabilization of system (1) can be achieved.

Let us introduce a definition which will be used in subsequent main results.

- 1 There exist two positive numbers  $k$  and  $b$ , such that  $\|x(t)\| \leq k \cdot e^{-bt}$ ,  $\forall t \geq 0$ .
- 2 There exists a positive number  $t_c$  such that  $x(t) = 0$ ,  $\forall t \geq t_c$ .

**Definition 1**

The system (1) is said to realize the generalized stabilization, provided that there exist a suitable control  $u$  such that the conditons (i) and (ii) are satisfied. In this case, the positive number  $b$  is called the exponential convergence rate and the positive number  $t_c$  is called the critical time.

Now we present the main result for the generalized stabilization of the system (1) via time-domain approach with differential inequalities.

**Theorem 1**

The system (1) realizes the generalized stabilization under the following control

$$u_1(t) = -(a_1 + b)x_1(t) - a_2x_2(t) - ax_1^{2\alpha-1}(t), \tag{2a}$$

$$\begin{aligned} u_2(t) = & -a_3x_1(t) - a_4x_1(t)x_3(t) - bx_2(t) \\ & - ax_2^{2\alpha-1}(t), \end{aligned} \tag{2b}$$

$$\begin{aligned} u_3(t) = & -a_5x_1(t) - a_6x_2(t) - (a_7 + b)x_3(t) \\ & - a_8x_1^2(t) - ax_3^{2\alpha-1}(t), \end{aligned} \tag{2c}$$

where  $a > 0, b > 0$ ,  $\alpha := \frac{p+q-1}{2p-1}$ , with  $p, q \in N$  and  $p > q$

In this case, the pre-specified exponential convergence rate and the guaranteed critical time are given by  $b$  and

$$t_c = - \frac{\ln \left[ \frac{\frac{a}{b}}{[x_1^2(0) + x_2^2(0) + x_3^2(0)]^{1-\alpha} + \frac{a}{b}} \right]}{-2(1-\alpha)b}, \tag{3}$$

respectively

**Proof.** From (1)-(2), the feedback-controlled system can be performed

$$\dot{x}_1 = -bx_1 - a(x_1)^{2\alpha-1}, \tag{4a}$$

$$\dot{x}_2 = -bx_2 - a(x_2)^{2\alpha-1}, \tag{4a}$$

$$\dot{x}_3 = -bx_3 - a(x_3)^{2\alpha-1}, \tag{4a}$$

Let

$$W(x(t)) = x^T(t)x(t). \tag{5}$$

The time derivative of  $W(x(t))$  along the trajectories of feedback-controlled system is given by

$$\begin{aligned} \dot{W} & = 2x_1 \cdot \dot{x}_1 + 2x_2 \cdot \dot{x}_2 \\ & = -2b \cdot x_1^2 - 2b \cdot x_2^2 - 2a \cdot x_1^{2\alpha} - 2a \cdot x_2^{2\alpha} \\ & = -2b \cdot W - 2a(x_1^{2\alpha} + x_2^{2\alpha}) \\ & \leq -2b \cdot W - 2a \cdot W^\alpha, \quad \forall t \geq 0. \end{aligned}$$

It follows that

$$\begin{aligned} (1-\alpha)W^{-\alpha}\dot{W} + 2(1-\alpha)bW^{1-\alpha} \\ \leq -2a(1-\alpha), \quad \forall t \geq 0. \end{aligned} \tag{6}$$

Define

$$Q(t) := W(x(t))^{1-\alpha}, \quad \forall t \geq 0. \tag{7}$$

From (6) and (7), it can be readily obtained that

$$\dot{Q} + 2(1-\alpha)bQ \leq -2a(1-\alpha), \quad \forall t \geq 0.$$

It is easy to deduce that

$$\begin{aligned} e^{2(1-\alpha)bt} \cdot \dot{Q}(t) + e^{2(1-\alpha)bt} \cdot 2(1-\alpha)bQ(t) \\ = \frac{d}{dt} [e^{2(1-\alpha)bt} \cdot Q(t)] \\ \leq -2a(1-\alpha)e^{2(1-\alpha)bt}, \quad \forall t \geq 0. \end{aligned}$$

It follows that

$$\begin{aligned} \int_0^t \frac{d}{dt} [e^{2(1-\alpha)bt} \cdot Q(t)] dt \\ = e^{2(1-\alpha)bt} \cdot Q(t) - Q(0) \end{aligned}$$

$$\leq \int_0^t -2a(1-\alpha)e^{2(1-\alpha)bt} dt$$

$$= \frac{-a}{b} (e^{2(1-\alpha)bt} - 1), \quad \forall t \geq 0.$$

Consequently, we have

$$Q(t) \leq \left[ Q(0) + \frac{a}{b} \right] \cdot e^{-2(1-\alpha)bt} - \frac{a}{b}, \quad (8)$$

$$\forall t \geq 0.$$

Hence, from (6), (7), and (8), we have

(i) If  $0 \leq t \leq t_c$ ,

$$W(x(t)) \leq \left[ \left( \|x(0)\|^{2-2\alpha} + \frac{a}{b} \right) \cdot e^{-2(1-\alpha)bt} - \frac{a}{b} \right]^{1/(1-\alpha)}; \quad \text{(ii) If } t \geq t_c,$$

$$x(t) = 0.$$

Consequently, we conclude that

(i) If  $0 \leq t \leq t_c$ ,

$$\|x(t)\| \leq \left[ \|x(0)\|^{2-2\alpha} + \frac{a}{b} \right]^{1/(2-2\alpha)} \cdot e^{-bt},$$

(ii) If  $t \geq t_c$ ,  $x(t) = 0$ ,

in view of (5) with above condition (i). This completes the proof.  $\square$

### 3. NUMERICAL SIMULATIONS

Consider the generalized Liu chaotic system of (1) with  $a_1 = -a_2 = 10$ ,  $a_3 = 40$ ,  $a_4 = -1$ ,  $a_5 = a_6 = 0$ ,  $a_7 = -2.5$ ,  $a_8 = 4$ , and  $x(0) = [4 \ 2 \ -2]^T$ . Our objective, in this example, is to design a feedback control such that the system (1) realize the generalized stabilization with the guaranteed exponential convergence rate  $b = 0.5$ . From (2), with  $a = 100$ ,  $p = 3$ ,  $q = 2$ , we deduce  $\alpha = 0.8$ ,

$$u_1(t) = -10.5x_1(t) + 10x_2(t) - 100x_1^{0.6}(t), \quad (9a)$$

$$u_2(t) = -40x_1(t) + x_1(t)x_3(t) - 0.5x_2(t) - 100x_2^{0.6}(t), \quad (9b)$$

$$u_3(t) = 2x_3(t) - 4x_1^2(t) - 100x_3^{0.6}(t), \quad (9c)$$

Consequently, by Theorem 1, we conclude that the system (1) achieves generalized stabilization with parameters of  $a_1 = -a_2 = 10$ ,  $a_3 = 40$ ,  $a_4 = -1$ ,  $a_5 = a_6 = 0$ ,  $a_7 = -2.5$ ,  $a_8 = 4$ , and feedback control law of (9). Furthermore, the exponential convergence rate and the guaranteed critical time are given by  $b = 0.5$  and  $t_c = 0.047$ .

The typical state trajectories of uncontrolled systems and controlled systems are depicted in Figure 1 and Figure 2, respectively. From the foregoing simulations

results, it is seen that the dynamic system of (1) achieves the generalized stabilization under the control law of (9).

### 4. CONCLUSION

In this paper, the concept of generalized stabilization for nonlinear systems has been introduced and the stabilization of generalized Liu chaotic control system has been studied. Based on the time-domain approach with differential inequalities, a suitable control has been presented such that the generalized stabilization for a class of Liu chaotic system can be achieved. Besides, not only the guaranteed exponential convergence rate can be arbitrarily pre-specified but also the critical time can be correctly estimated. Finally, some numerical simulations have been offered to show the feasibility and effectiveness of the obtained results.

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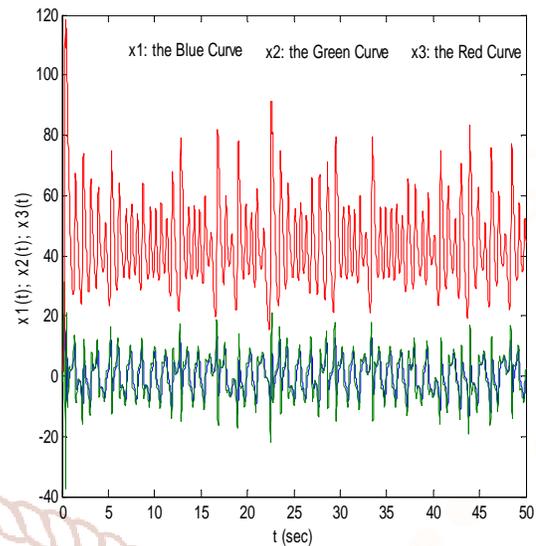
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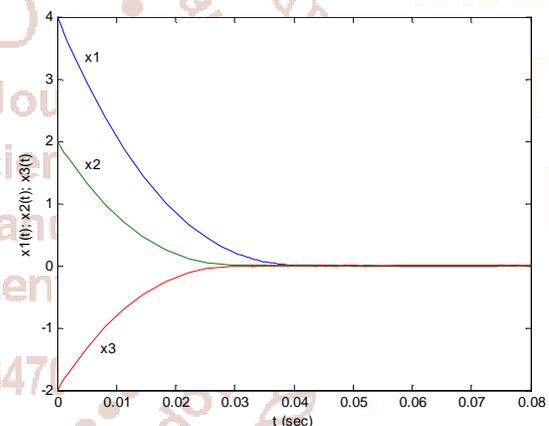
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**Figure 1:** Typical state trajectories of the system (1) with  $u=0$ ,  $a_1=-a_2=10$ ,  $a_3=40$ ,  $a_4=-1$ ,  $a_5=a_6=0$ ,  $a_7=-2.5$ , and  $a_8=4$ .



**Figure 2:** Typical state trajectories of the feedback-controlled system of (1) with (9).